

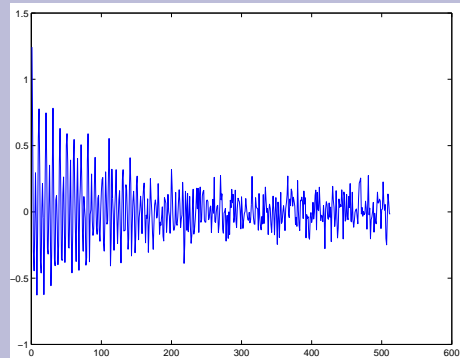
# Advanced NMR Processing

Ulrich Günther

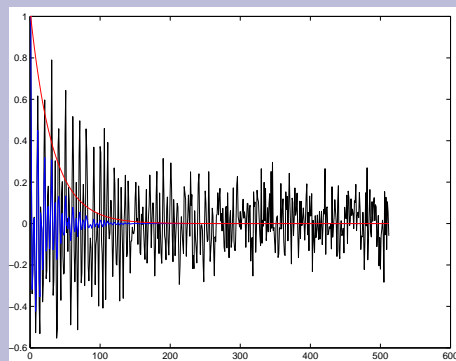
EuroLabCourse "Advanced Computing in  
NMR Spectroscopy", Florence, Sept. 2001

# NMR Processing

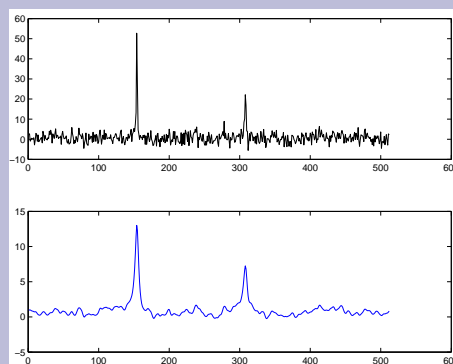
Free induction decay (FID)



Apodization



FFT



# NMR Processing

- **Main processing steps**
  - Apodization
    - \* noise reduction
    - \* resolution enhancement
  - On-resonance water suppression
  - Zero-filling / linear prediction
  - Fourier transform
- **Post-processing**
  - Baseline correction
  - Denoising

## FT - Convolution properties

A convolution is defined as

$$f_1(t) \star f_2(t) = \int_{-\infty}^{+\infty} f_1(\tau) f_2(t - \tau) d\tau$$

$$\mathcal{F}\{f_1(t) \star f_2(t)\} = \mathcal{F}\{f_1(t)\} \cdot \mathcal{F}\{f_2(t)\}$$

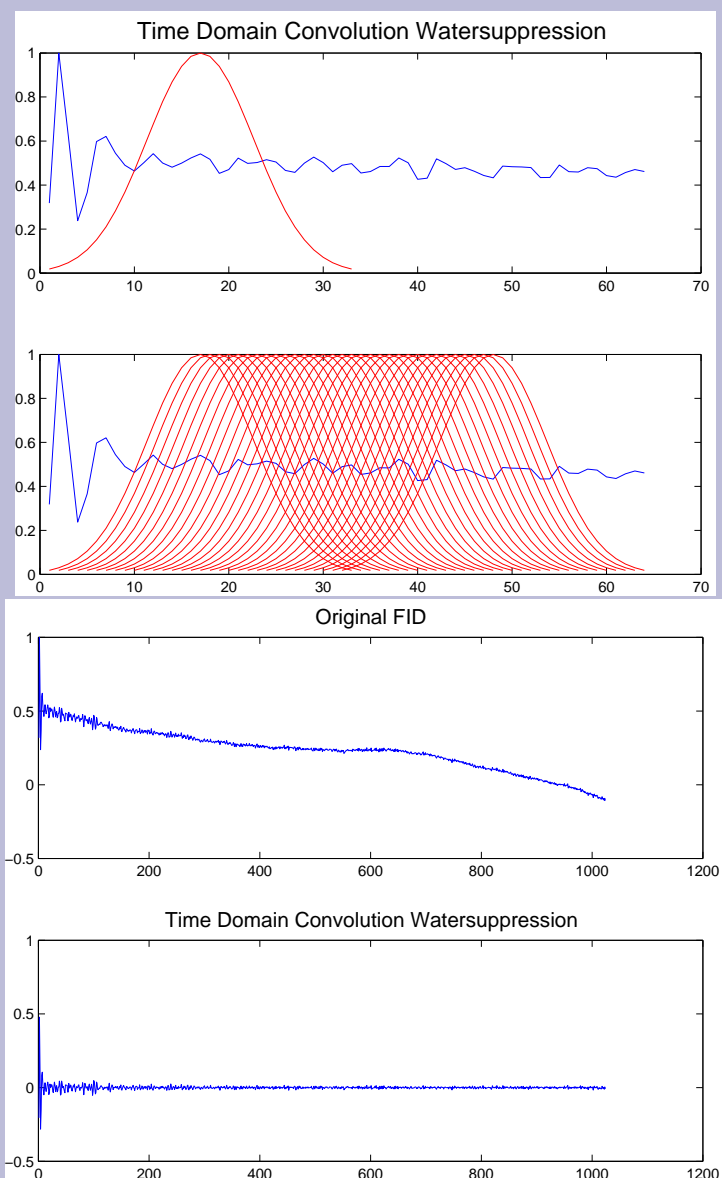
For discrete data this is equivalent to

$$DFT(\mathbf{da})_n = \frac{1}{\sqrt{N}} \sum_{j=0}^{N-1} f_j A_{n-j}$$

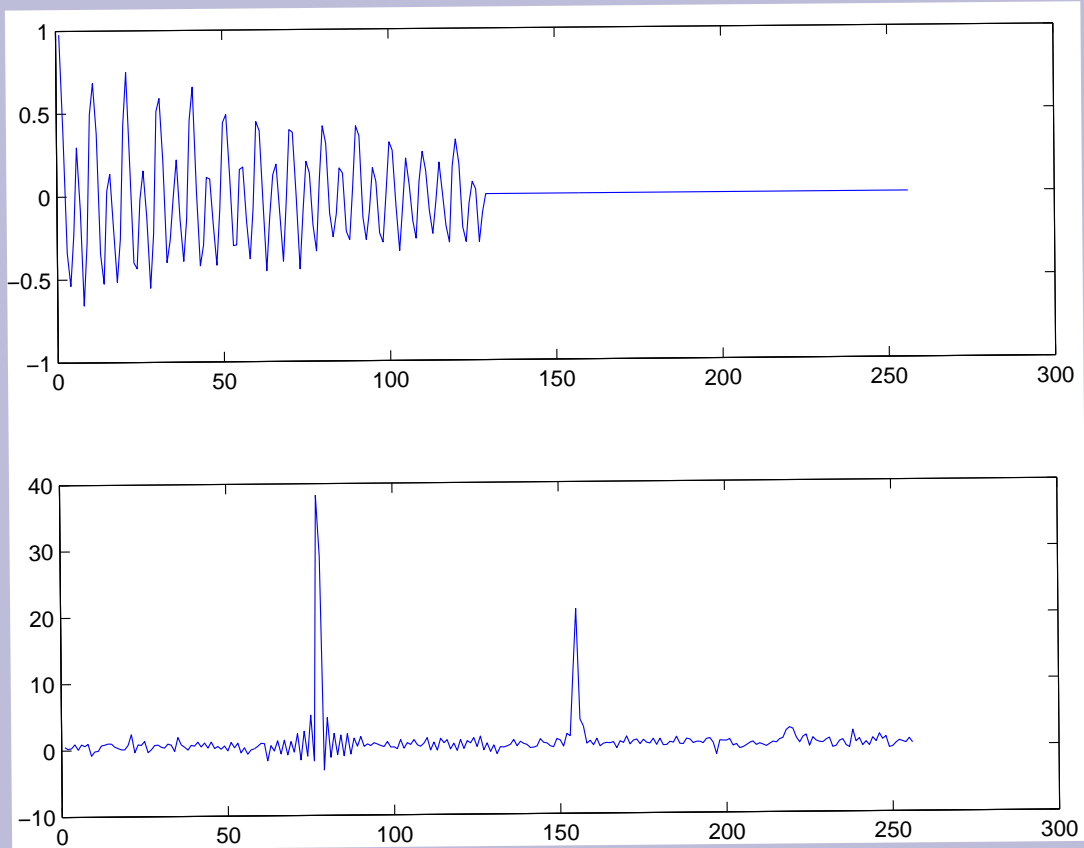
Special case: Convolution of a function with itself

$$\mathcal{F}\{f_1(t) \star f_1(t)\} = \mathcal{F}\{f_1(t)\} \cdot \mathcal{F}\{f_1(t)\}$$

# Water Suppression by Time-Domain Convolution

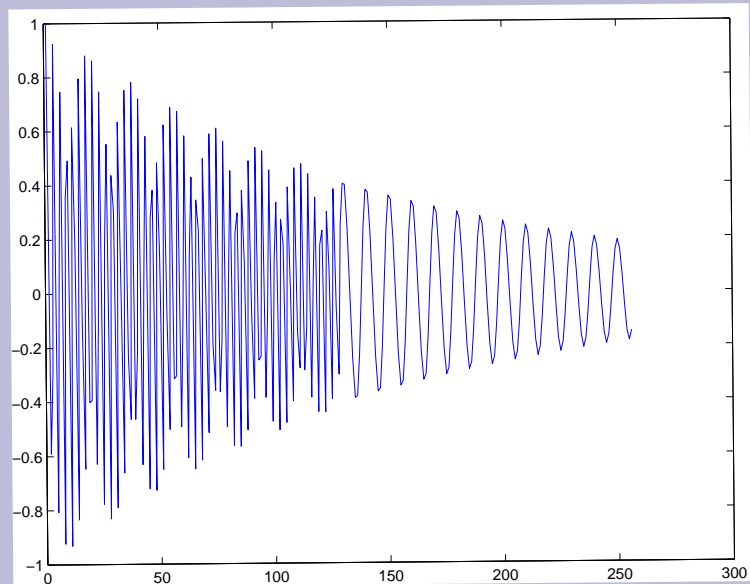


# Zero-filling as extrapolation



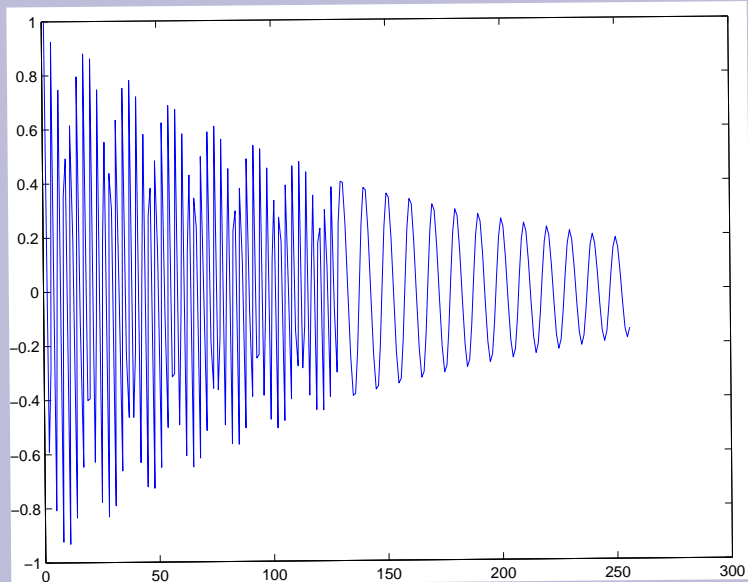
Zero-filling may cause truncation artefacts

# Fourier transform properties

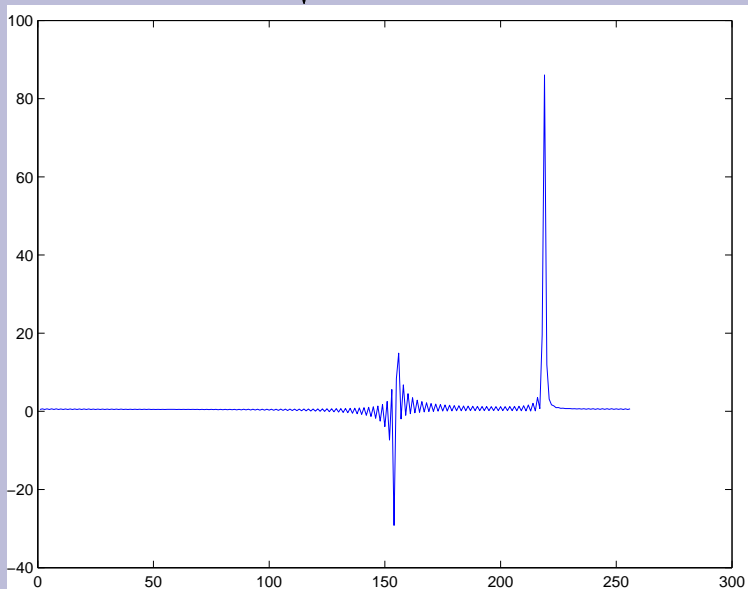


↓ FFT  
?

# Fourier transform properties

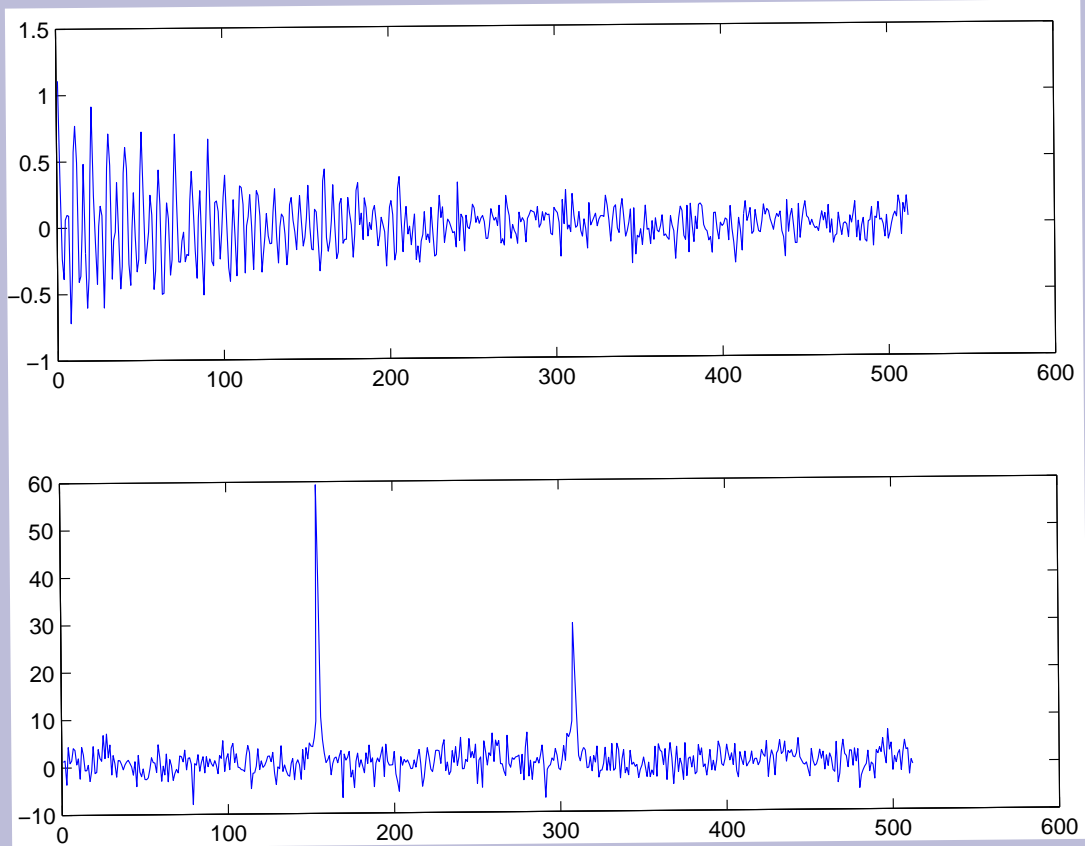


↓ FFT





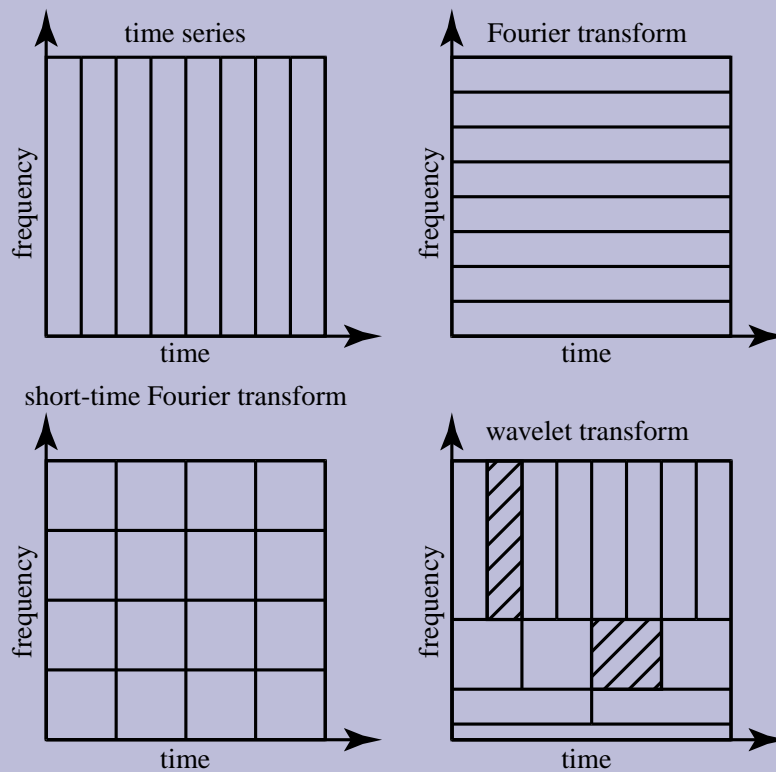
# FFT and noisy data



# Wavelet Transforms

- *Ondelettes* (wavelets) - Yves Meyer, 1980s
- *Wavelet*  $\leftarrow$  “waving” above and below the  $x$ -axis  $\Leftrightarrow$  integrate to zero.
- Wavelets **chop up data into frequency components**, and **analyze each frequency component with a resolution matched to its scale**.
- Applications:
  - data approximation (smoothing)
  - noise reduction
  - data compression (jpeg, mpeg, mp3)
  - time-frequency analysis
  - image analysis
- Fast dyadic wavelet transform (DWT) -  $N$  operations, compare: FFT  $N \log(N)$ .

# Wavelet Transforms



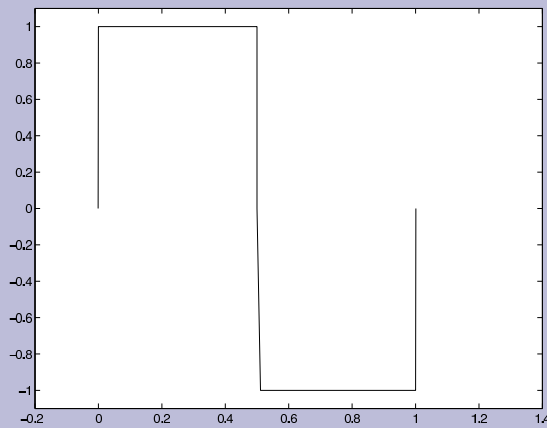
- **FT:** high resolution in the time domain, each point contains information about all frequencies.
- **WFT:** divides the time-frequency plane in rectangular boxes. The resolution in time is increased at the expense of the frequency resolution.
- **DWT:** scaling of the basis functions relative to their support.

# The Haar System

A. Haar in 1909, appendix to a thesis.

The Haar function is a simple step function

$$\psi(x) = \begin{cases} 1, & x \in [0, \frac{1}{2}[ \\ -1, & x \in [\frac{1}{2}, 1[ \end{cases}$$

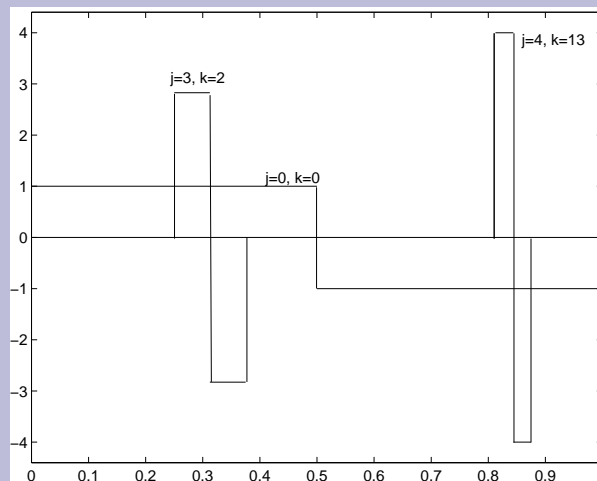


# Dilatations and Translations

The Haar function  $\psi$  is used to define a *mother wavelet*.  
*Dilatations and translations*  $\leftarrow$  other wavelets.

$$\psi_{j,k}(x) = 2^{j/2} \psi(2^j x - k). \quad (1)$$

Integer translation indices  $k$  and dilatation indices  $j$ .



## Properties of the Haar wavelet

Haar wavelets  $\psi_{j,k}$  have a **compact support**

$$\text{supp}(\psi_{j,k}) = [k2^{-j}, (k+1)2^{-j}[ \quad (2)$$

i.e. they are zero outside this interval.

For each Haar wavelet  $\psi_{j,k}$  the integral

$$\int_{-\infty}^{\infty} \psi_{j,k}(x) dx = 0.$$

I.e. the area above the  $x$ -axis is equal to the area below the  $x$ -axis.

# Orthonormal System

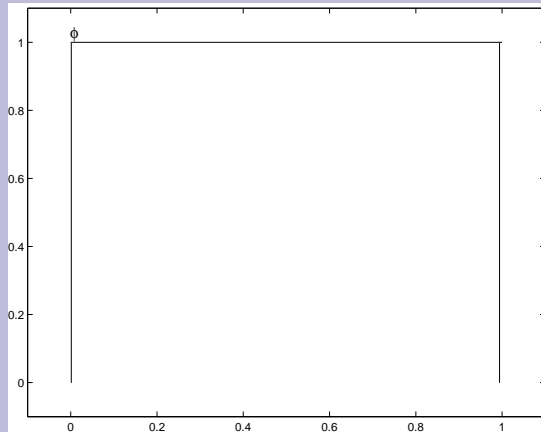
$\{\psi_{j,k}, j, k \in \mathbb{Z}\}$  constitutes a **complete orthonormal basis in  $L_2$**

$\iff$  **Any square integrable function can be approximated arbitrarily well by a linear combination of these basis functions.**

$L_2(\mathbb{R})$  = space of complex valued functions with a finite  $L_2$ -norm:

$$\|F\|_2 = \left( \int_{-\infty}^{\infty} |f(x)|^2 dx \right)^{1/2}$$

# Father Wavelet



- For the Haar wavelet the father wavelet is a simple step function:

$$\phi = \begin{cases} 1, & x \in [0, 1[ \\ 0, & x \notin [0, 1[ \end{cases} .$$

- Required for fast DWT and wavelet construction (father wavelet  $\rightarrow$  mother wavelet).
- With the scaling function  $\phi$  we can expand our original set to  $\{\phi_{j_0,k}, \psi_{j,k}, j \geq j_0, k \in \mathbb{Z}\}$ . The combined set is again an orthonormal basis in  $L_2$ .



# Signal Approximation in the Haar System

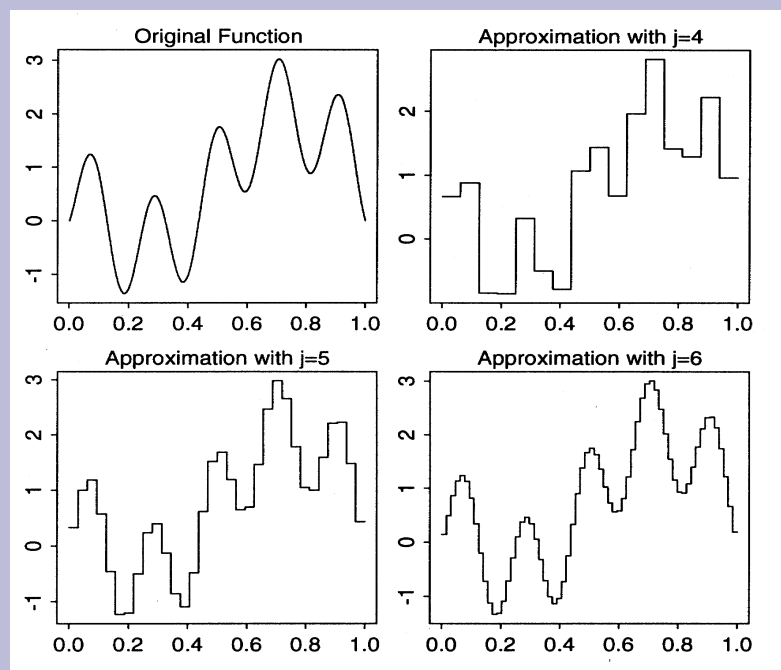
A signal can be represented as

$$f(x) = c_{00}\phi(x) + \sum_{j=0}^{n-1} \sum_{k=0}^{2^j-1} c_{j,k}\psi_{j,k}(x) \quad (3)$$

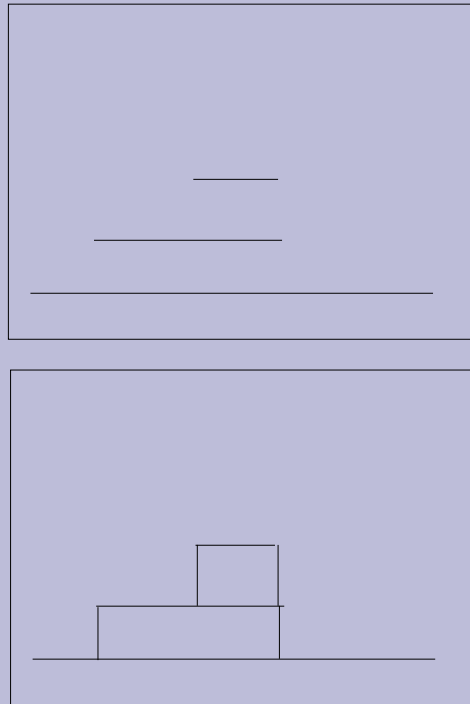
$c_{j,k}$  are wavelet coefficients,

$\psi_{j,k}$  are wavelets derived from a mother wavelet  $\psi$

$\phi$  is a scaling function (father wavelet)



## Step Function $\rightarrow$ Wavelet Transform



Approximations of a function with increasing resolution by a step function.

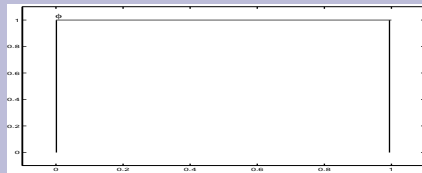
- Any  $L_2$ -function can be approximated by a simple step function.
- The approximation converges for infinitely small steps.
- Step functions do not constitute an orthonormal basis for  $L_2$ .

# Multiresolution Analysis (MRA)

- G. Mallat
- general framework to construct wavelet bases
- start from a father wavelet
- derive an *orthonormal* mother wavelet
- and wavelet subspaces suitable to approximate function with increasing resolution

# MRA 1

Start with the scaling function (father wavelet)



$$\phi_{0,k} = \phi(x - k)$$

Set  $\{\phi_{o,k}\}$  = orthonormal basis for a reference space  $V_0$ .

The functions in  $V_0$  have the form

$$f(x) = \sum_k c_k \phi(x - k)$$

Define a function space  $V_0$ , functions in  $[k, k + 1[$

$$V_0 = \left\{ f(x) = \sum_k c_k \phi(x - k) \right\}.$$

## MRA 2

Starting from  $V_0$  we define linear spaces

$$V_1 = \{h(x) = f(2x) : f \in V_0\}$$

⋮

$$V_j = \{h(x) = f(2^j x) : f \in V_0\}$$

$V_1$ : all functions constant on  $[\frac{k}{2}, \frac{k+1}{2}[$ .

Set  $\{\phi_{1,k}\}$  is an orthonormal basis in  $V_1$

with

$$\phi_{1,k}(x) = \sqrt{2}\phi(2x - k)$$

.

## MRA 3

$$V_0 \longrightarrow V_1 \longrightarrow V_j$$

Basis functions of  $V_j$  are  $\phi_{j,k} = 2^{j/2}\phi(2^jx - k)$ .

$\phi$  generates a sequence of **spaces**  $\{V_j, j \in \mathbb{Z}\}$  **which are nested:**

$$V_0 \subset V_1 \subset \dots \subset V_j \subset \dots$$

$$V_j \subset V_{j+1}, j \in \mathbb{Z}.$$

If in addition every square integrable function can be approximated by functions in

$$\bigcup_{j \geq 0} V_j$$

than  $\{V_j, j \in \mathbb{Z}\}$  is a MRA!

## MRA 4

$\{V_j, j \in \mathbb{Z}\}$  is not a basis in  $L_2$ !

To obtain a basis we must **orthogonalize** it.

$\implies$  find a  $W_0$  for which

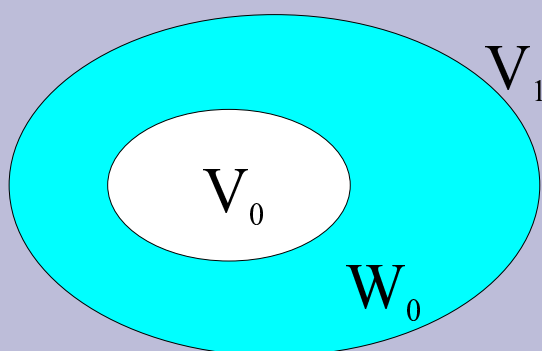
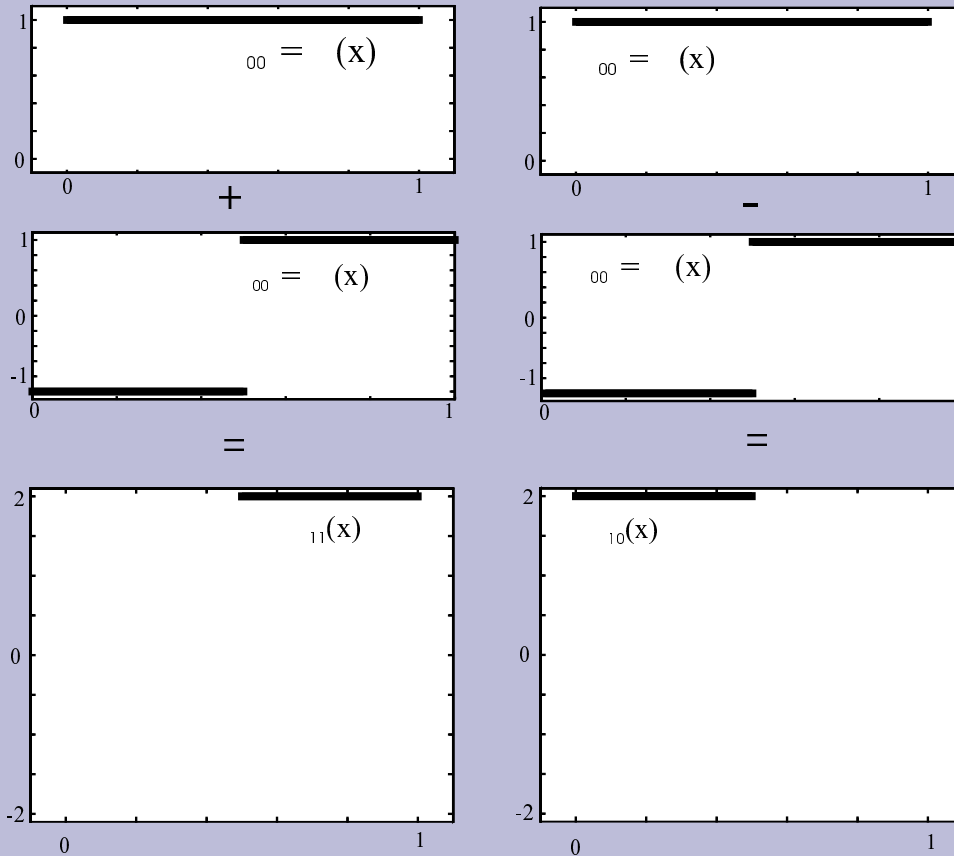


Figure 1:  $V_1 = V_0 \oplus W_0$ .

$$V_1 = V_0 \oplus W_0.$$

$W_0$  is the orthogonal complement of  $V_0$  in  $V_1$ .

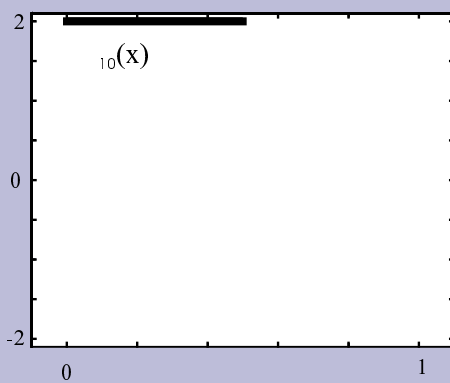
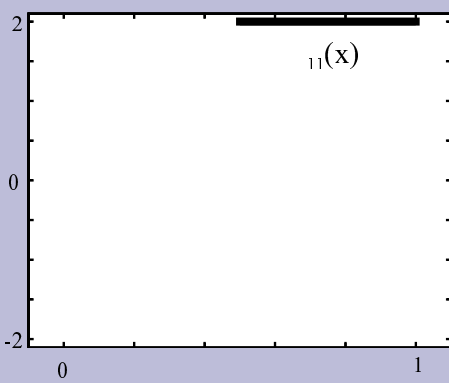
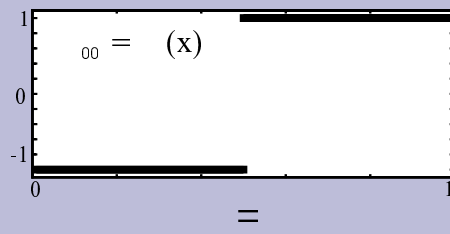
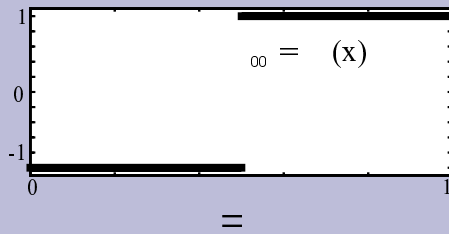
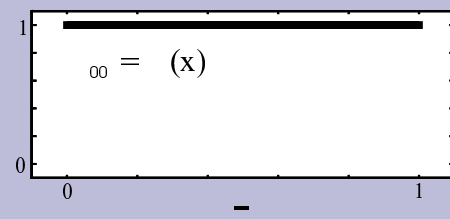
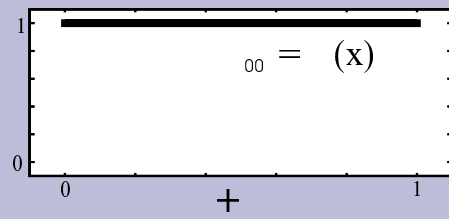
# MRA 6



$\phi_{1k}$  must be a linear combination of  $\phi_{0k}$  and  $\psi_{0k}$ :

$$\begin{aligned} \phi_{10}(x) &= \sqrt{2}\phi(2x) = \sqrt{2}\{\phi_{00}(x) - \psi_{00}(x)\}/2 \\ &= \frac{1}{\sqrt{2}}\{\phi_{00}(x) - \psi_{00}(x)\} \end{aligned}$$





$$\phi_{11}(x) = \sqrt{2}\phi(2x - 1) = \frac{1}{\sqrt{2}}\{\phi_{00}(x) + \psi_{00}(x)\}/2 .$$

## MRA 7

Repeat the same process for higher values of  $j$ .

—→ consecutive summation of subspaces

$$\begin{aligned}V_{j+1} &= V_j \oplus W_j \\ &= V_j \oplus W_{j-1} \oplus W_j \\ &= V_0 \oplus W_0 \oplus W_1 \oplus \dots \oplus W_j \\ &= V_0 \oplus \bigoplus_{l=0}^j W_l\end{aligned}$$

Finally this sum of nested spaces spans :

$$L_2(\mathbb{R}) = V_0 \oplus \bigoplus_{l=0}^j W_l.$$

$$L_2(\mathbb{R}) = V_0 \oplus \bigoplus_{l=0}^j W_l.$$

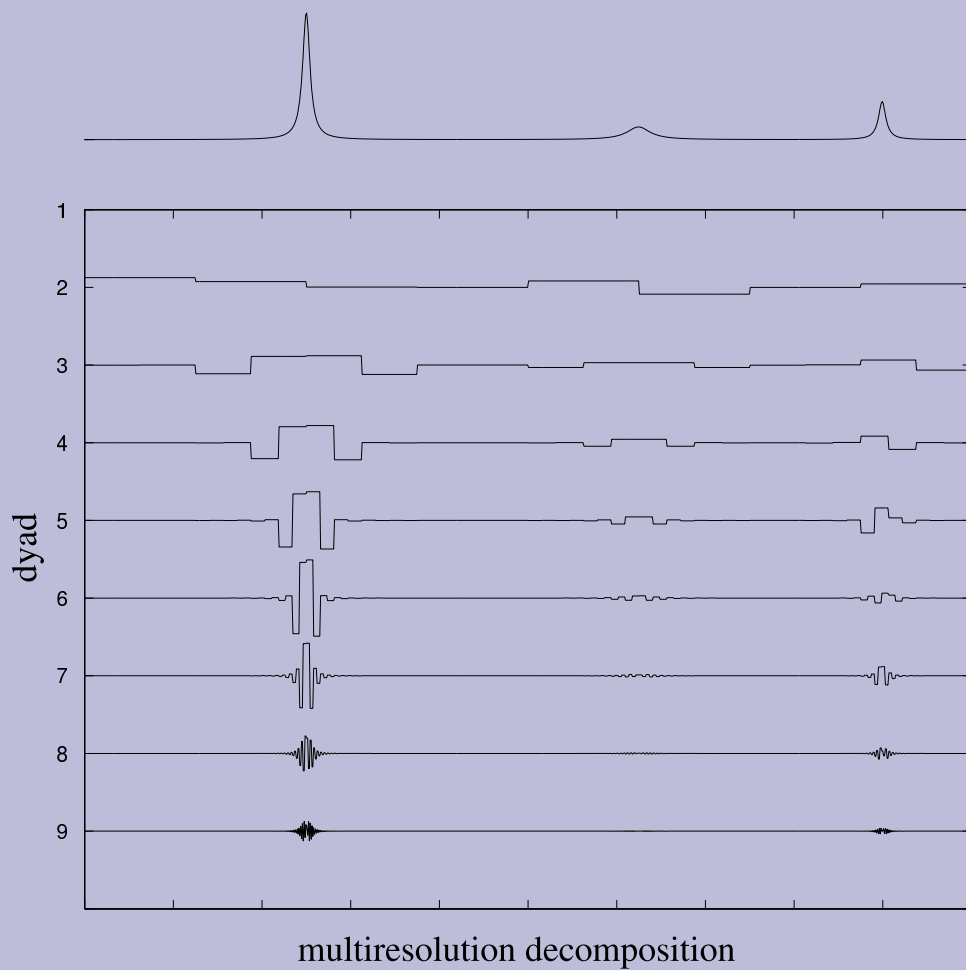
⇒ every square integrable function can be represented as the series

$$f(x) = \sum_k \alpha_{0,k} \phi_{0,k}(x) + \sum_{j=0}^{\infty} \sum_k \beta_{j,k} \psi_{j,k}(x).$$

- The father wavelet  $\phi$  creates a MRA
- $\alpha_{0,k}$  and  $\beta_{0,k}$  = coefficients for mother and father wavelets
- $f$  provides a location in time and frequency
  - The location in time is determined by  $k$
  - the location in frequency by  $j$  (the larger  $j$ , the higher the frequency related to  $\psi_{j,k}$ ).

$$f(x) = \alpha_{00}\phi(x) + \sum_{j=0}^{n-1} \sum_{k=0}^{2^j-1} c_{j,k}\psi_{j,k}(x), \quad (4)$$

Summation for  $f(x)$  stops at  $2^{j-1}$  for finite data sets  
 $N = 2^{j-1}$  =number of data points.



# Thresholding

Noise reduction = Search for the largest "true" wavelet coefficients.

Hard thresholding function: "keep or kill" selection:

$$\delta_{\text{hard}}(x) = \begin{cases} x, & \text{if } |x| > \lambda \\ 0, & \text{if } |x| < \lambda \end{cases}$$

Soft thresholding:

$$\delta_{\text{soft}}(x) = \begin{cases} x - \lambda, & \text{if } x > \lambda \\ 0, & \text{if } |x| \leq \lambda \\ x + \lambda, & \text{if } x < -\lambda \end{cases}$$

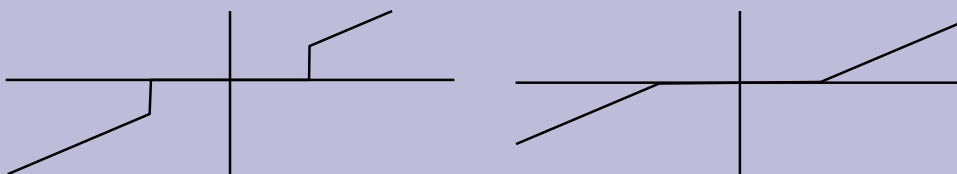
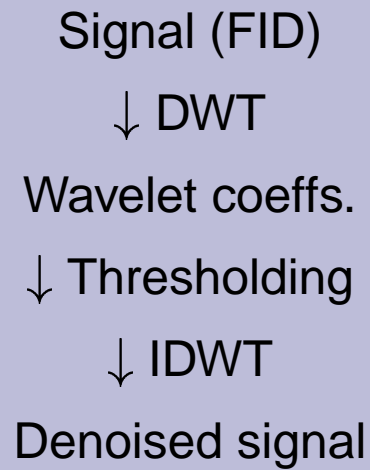


Figure 2: Hard and soft thresholding.

*Universal threshold* selection for white noise:

$$\lambda = \sigma \sqrt{2 \log n} / \sqrt{n}$$

# Denoising Procedure



## Smoother Wavelet Bases

Must form an orthonormal basis for  $L_2(\mathbb{R})$ .

Minimize the wavelet coefficients for smooth functions.

Number constraints during the design of wavelets determines number of vanishing moments.

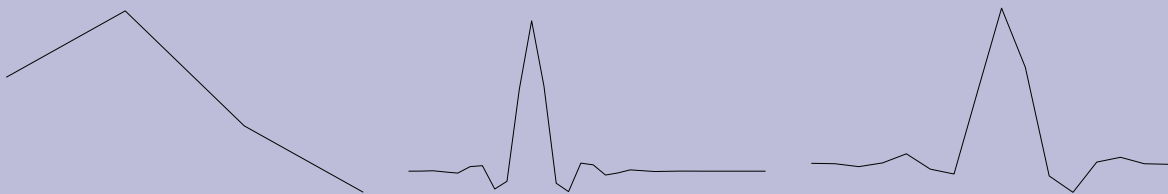
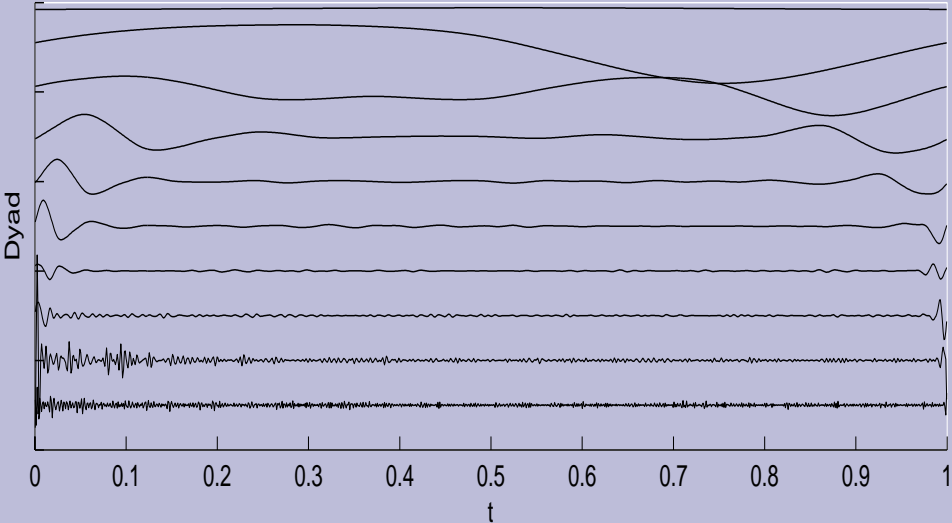


Figure 3: Types of wavelets from left to right: Daubechies (4), Coiflet (5) and Symmlet (8).

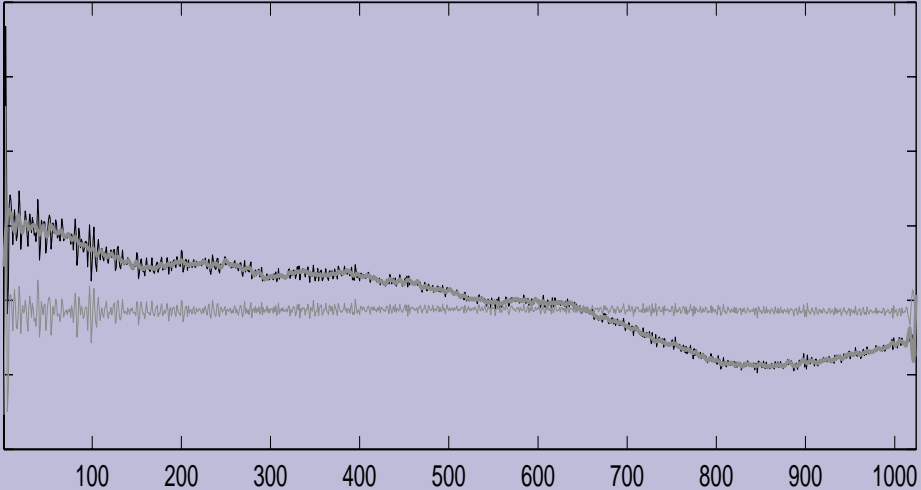
- Daubechies wavelets: vanishing moments for mother but not for father wavelets, asymmetric.
- Coiflets: additional vanishing moments for father wavelets.
- Symmlets are as close to symmetry as possible.

# WAVEWAT

Multiresolution Decomposition

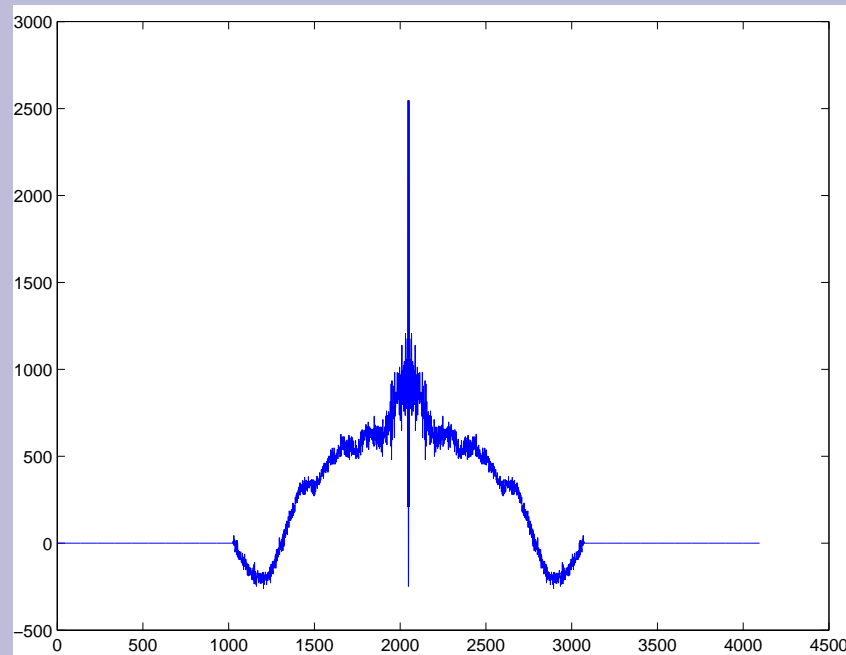


J=7 Reconstruction.



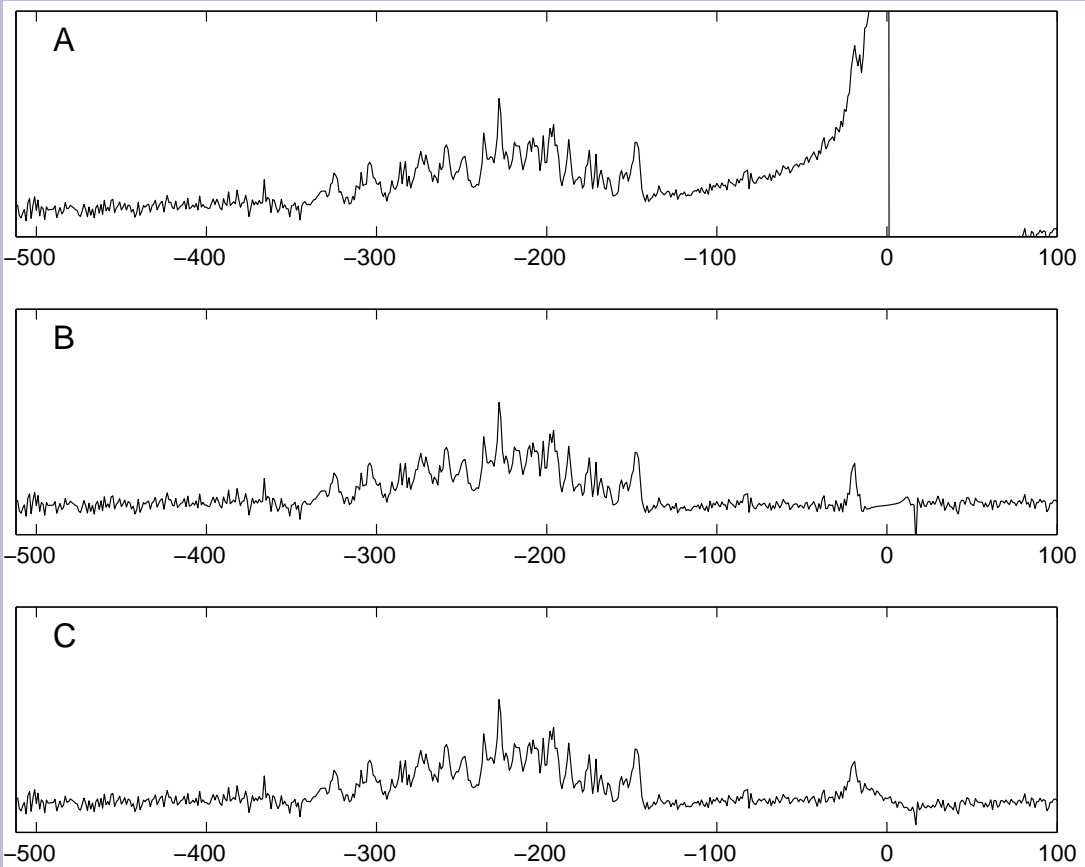


## Mirror reflections of FID to avoid edge effects

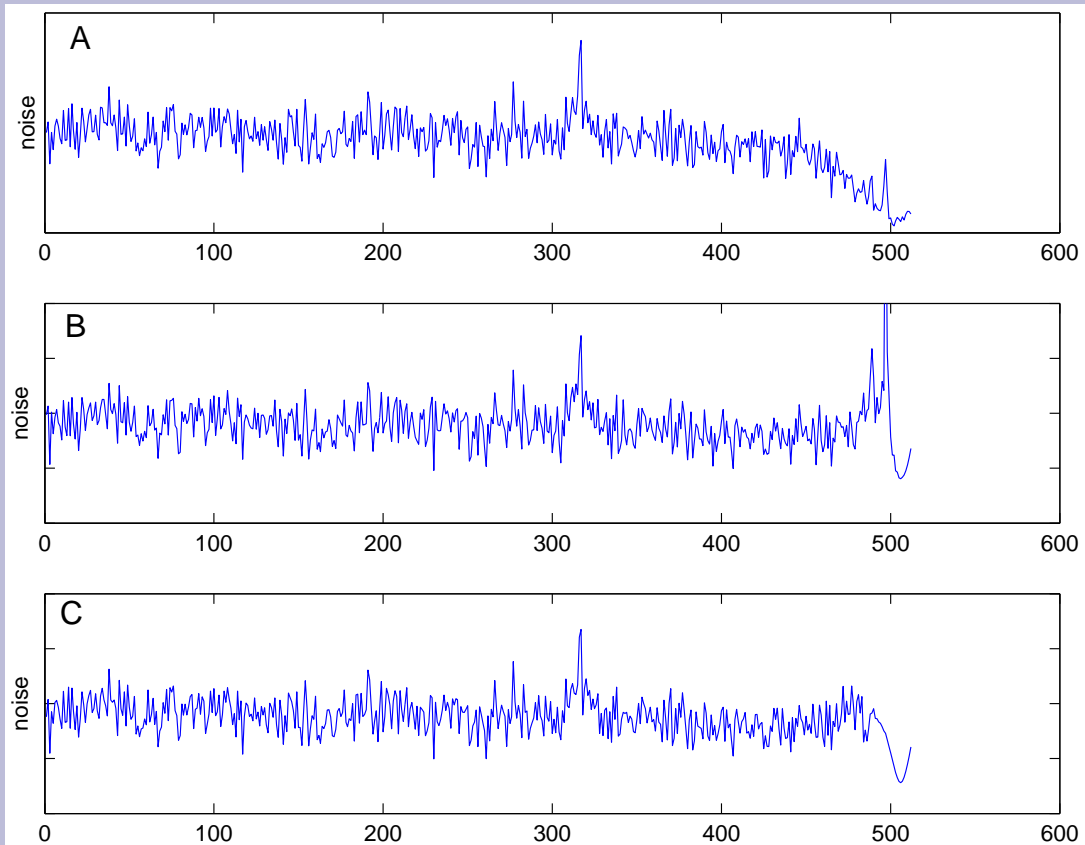


Use ZF to increase number of dyadic levels!

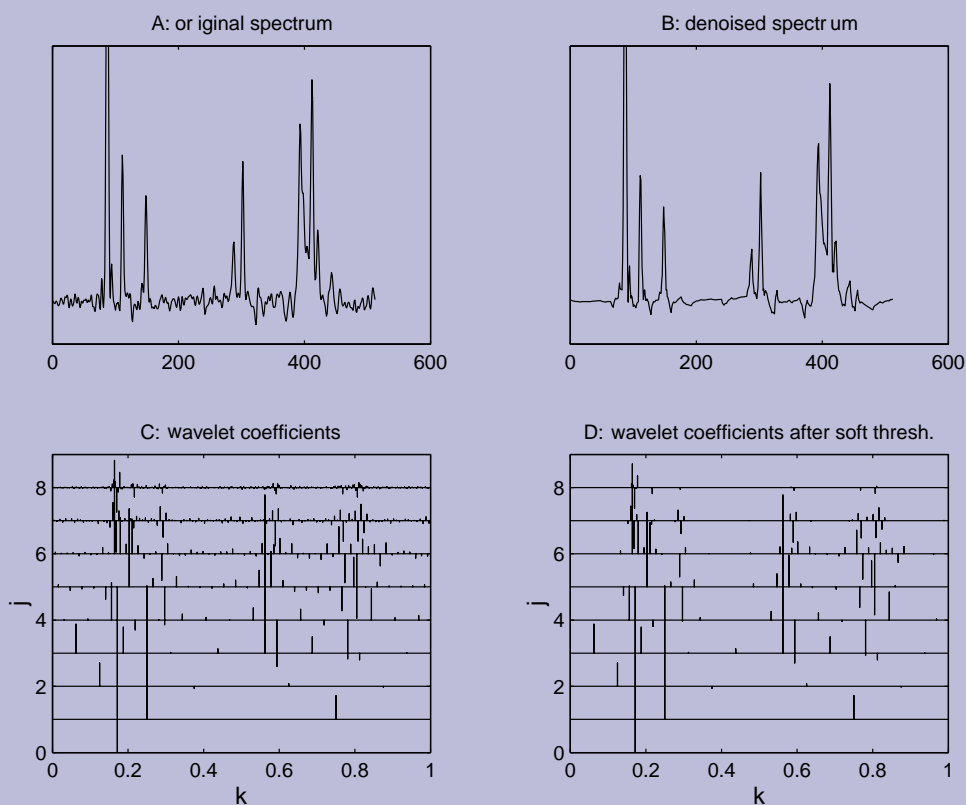
# WAVEWAT vs. TD Convolution Difference



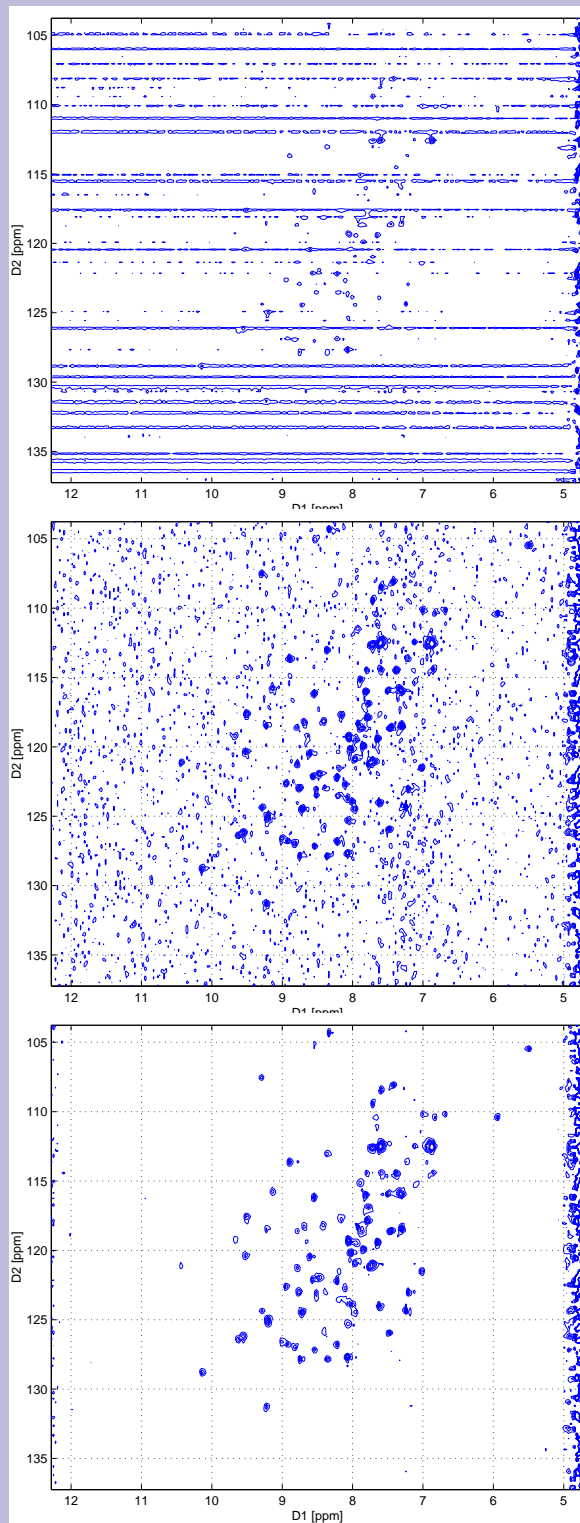
# Noise Levels

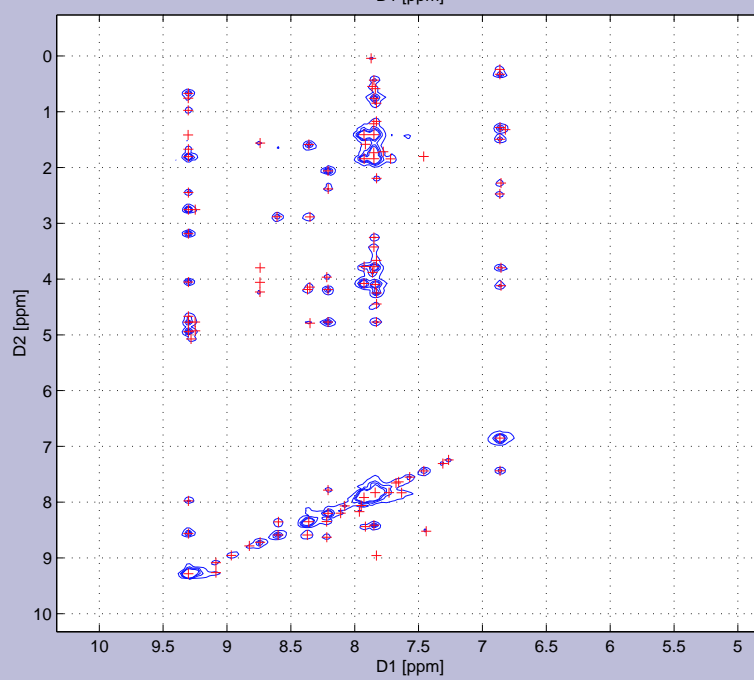
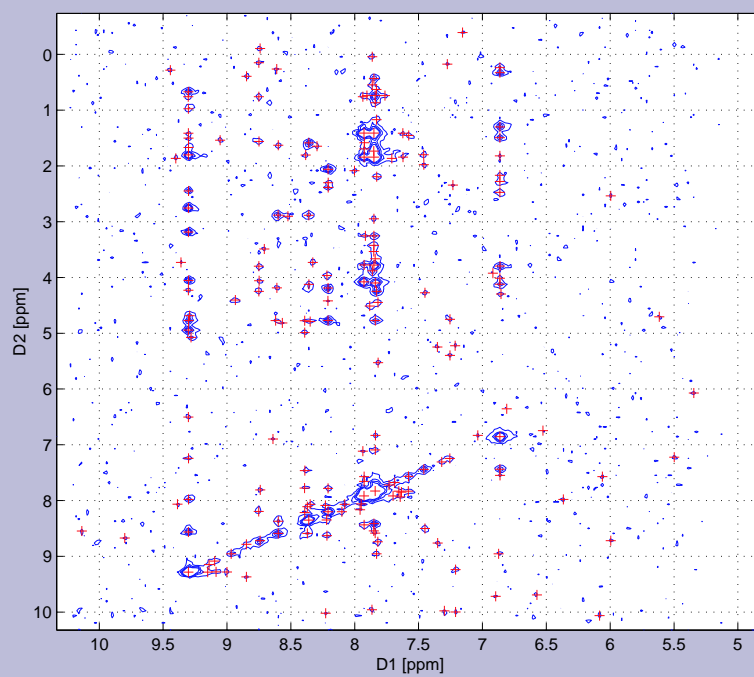


# Noise Suppression



Wavelet shrinkage using a Daubechies (4) wavelet and soft thresholding.





# SVD

$$\mathbf{H} = \mathbf{U}_{LL} \mathbf{S}_{LM} \mathbf{V}_{MM}^\dagger.$$

$\mathbf{U}$  and  $\mathbf{V}$  are  $(L \times L)$  and  $(M \times M)$  unitary matrices  
 $\mathbf{S}$  is  $(L \times M)$  in size and contains the singular values.

Singular values  $\iff$  signal components in  $\mathbf{H}$ .

Large signal components  $\iff$  large singular values.

## SVD noise suppression

FID:

$$f = (s_0, s_1, \dots, s_{N-1})$$

derive Hankel type matrix:

$$\mathbf{H} = \begin{pmatrix} s_0 & s_1 & s_2 & \cdots & s_{M-1} \\ s_1 & s_2 & s_3 & \cdots & s_M \\ \vdots & \vdots & \vdots & \cdots & \vdots \\ s_{L-1} & s_L & s_{L+1} & \cdots & s_{N-1} \end{pmatrix}.$$

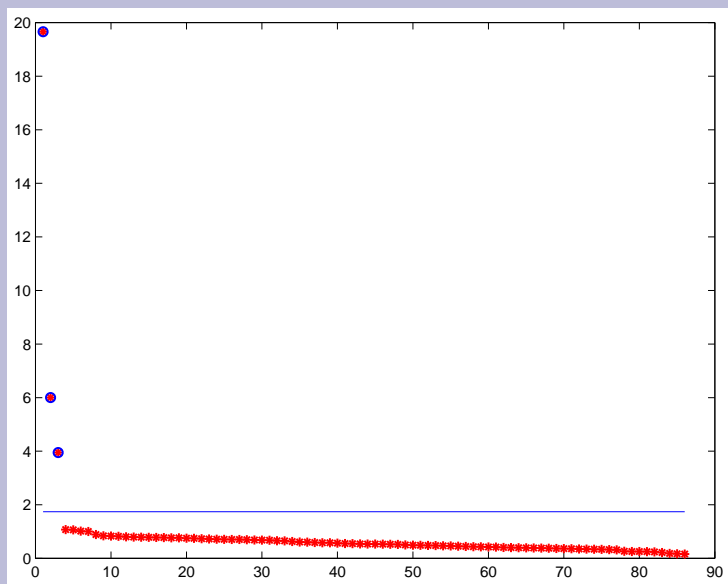
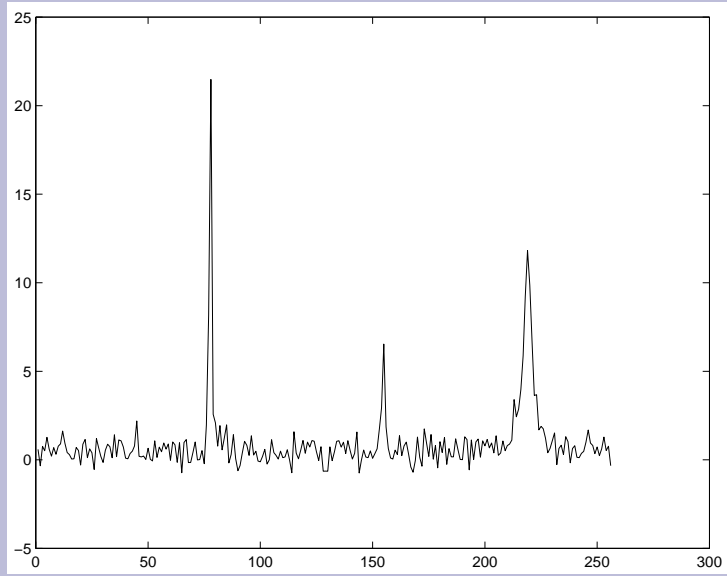
No noise: Rank of  $\mathbf{H}$  is equivalent to the number of signals in the FID.

Noise:  $\rightarrow$  full rank.

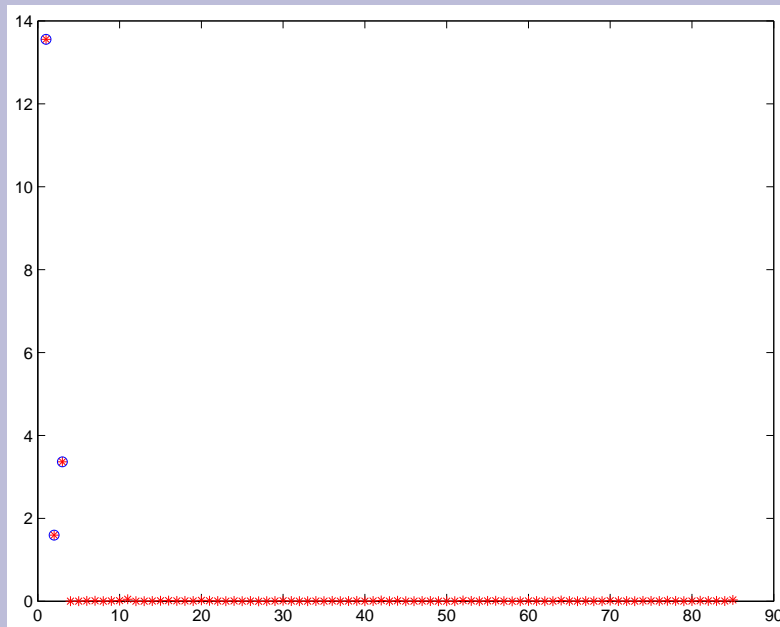
Periodic components in the FID will be represented singular values.

$O((L - M)^2 M)$  process!



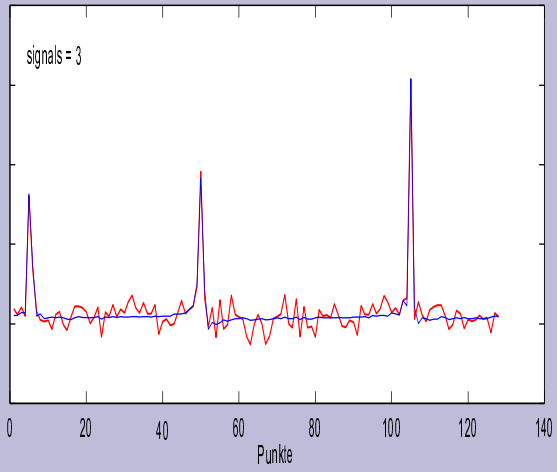


# Differential plot of the singular values

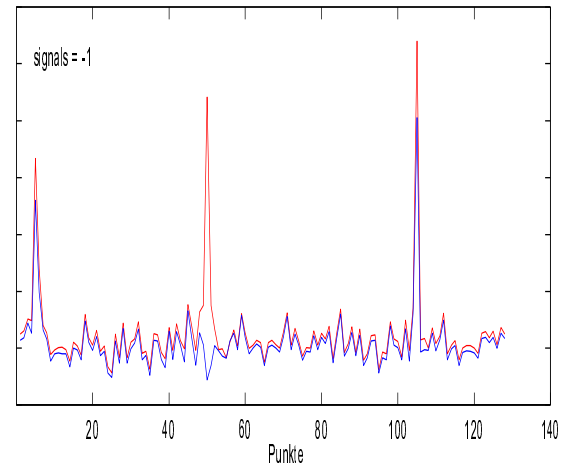


# SVD Applications

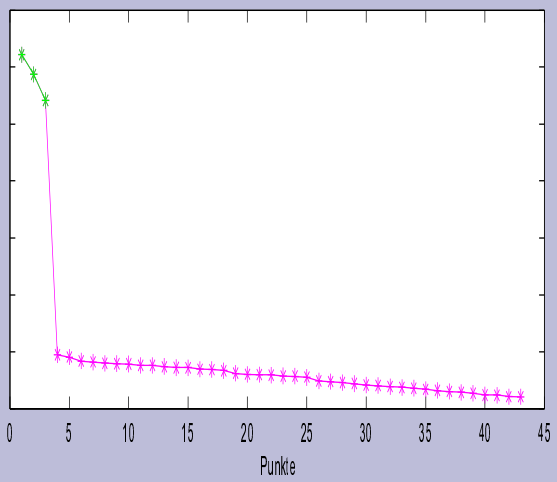
a



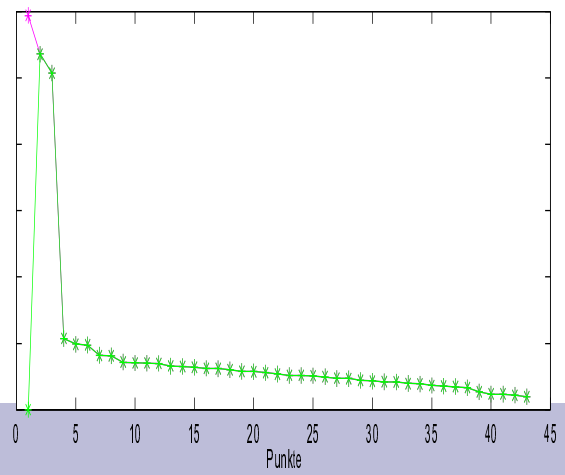
b



c



d



# Linear prediction

LP-SVD: by Kumaresan and Tufts modified by Porat and Friedlander.

Forward linear prediction: describe a data point by a linear combination of  $K$  preceding points:

$$x_n = \sum_{k=1}^K a_k x_{n-k}, \quad (5)$$

Backward linear prediction model by a linear combination of the  $K$  following points:

$$x_n = \sum_{k=1}^K b_k x_{n+k}. \quad (6)$$

$$x_n = \sum_{k=1}^K b_k x_{n+k}. \quad (7)$$

Equation 5 in matrix form:

$$x = a \cdot Y$$

inversion of  $Y \rightarrow$  matrix coefficients  $a$ :

by SVD:

$$Y = U \Lambda V^\dagger$$

$\rightarrow Y'$  is obtained by

$$Y^\dagger = V \Lambda^{-1} U^\dagger.$$

$$a = V \Lambda^{-1} U^\dagger x.$$

Prediction coefficients  $a_k \rightarrow$  frequencies and damping by calculating the roots of the polynomial

$$z^K - a_1 z^{K-1} - \dots - c_K = 0.$$

The  $z^K$  must be damped exponentials!

The signal  $x_n$  is described by

$$x_n = \sum_{k=1}^K A_k z_k^{n-1}. \quad (8)$$

$A_k$  = encode phase and amplitude of each signal determined by an additional SVD of  $\mathbf{Z}$ .

$$x = \mathbf{AZ}.$$

# Forward-backward LP

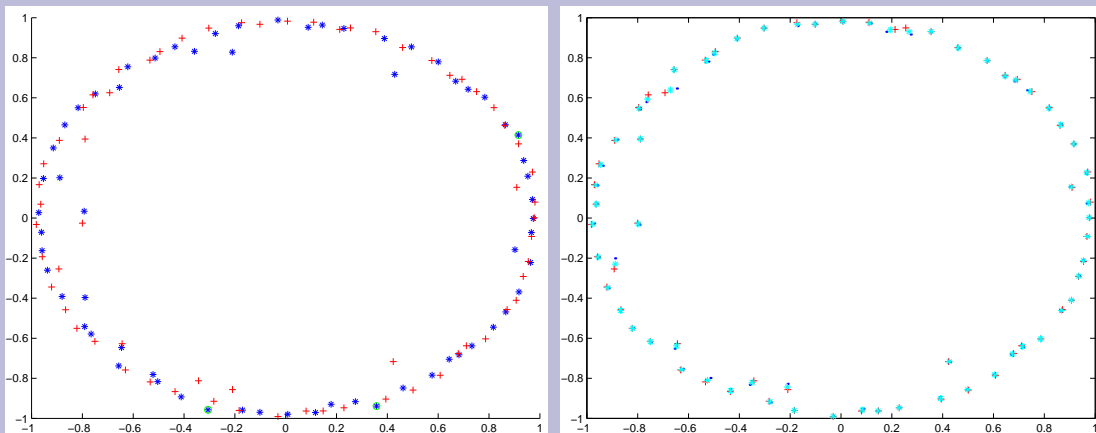
Calculate  $A_k$  LP coefficients for forward and backward LP

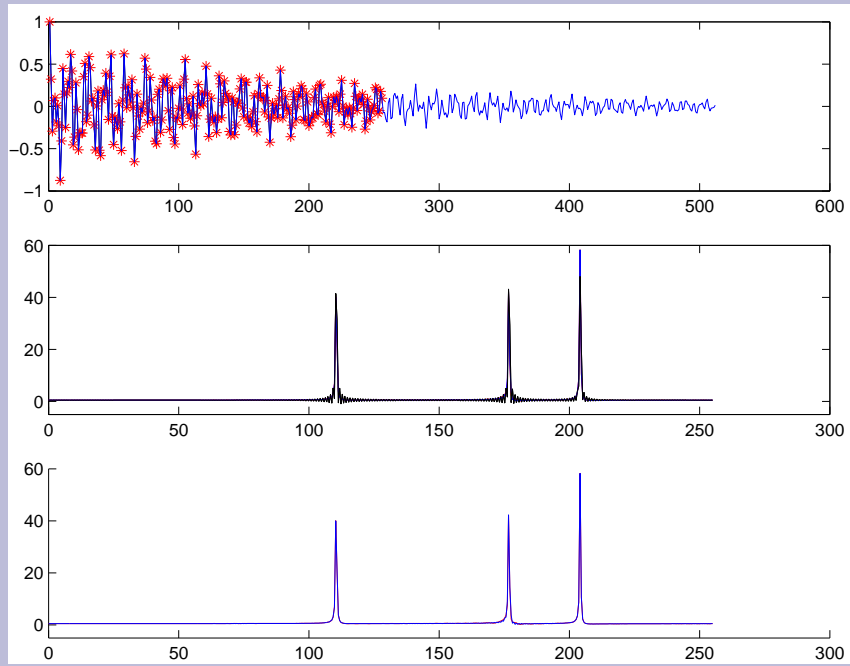
→ two sets of polynomial roots

Forward linear prediction: complex roots inside the unit circle, outside: noise / exp. incr. signal

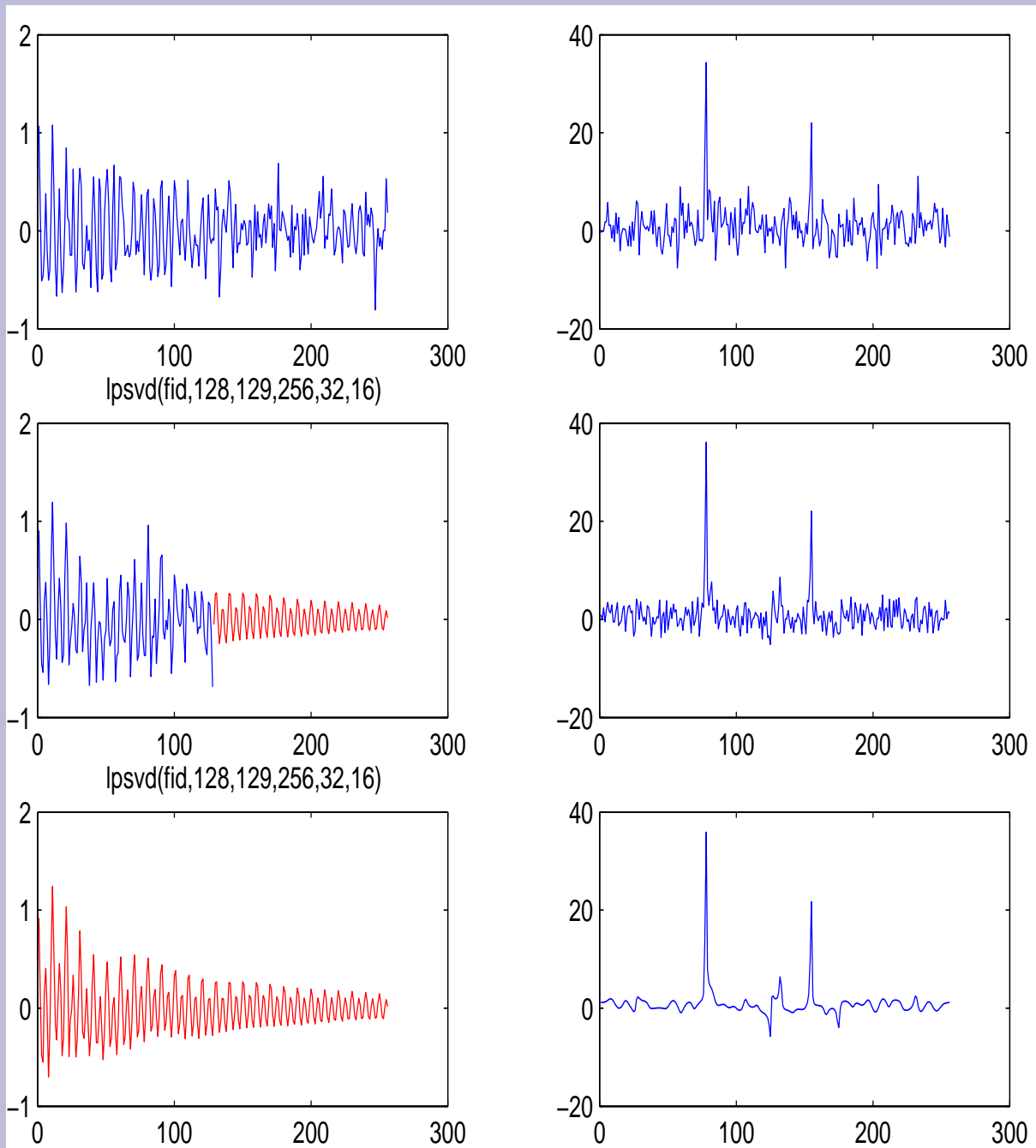
Backward linear prediction: inside the unit circle

Root reflection: replace  $z_k$  by  $z_k/|z_k|^2 \implies$  noise related signals are eliminated

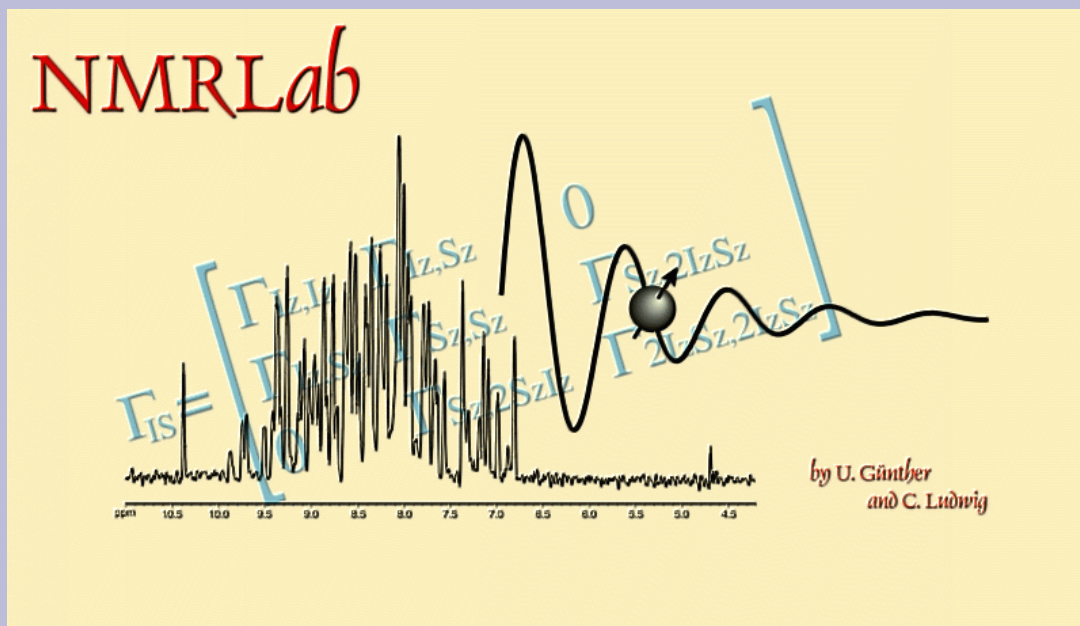








S/N enhancement by returning the complete predicted FID (`lpsvd(fid, 128, 1, 256, 32, 16)`)



- Standard processing: FFT, apodization functions, phase correction, baseline correction (FLATT and other algorithms), linear prediction (lpsvd with root reflection), Cadzow algorithm
- Export to NMRVIEW, AURELIA
- Peak picking: Export of peak lists to DYANA (NOAH)
- SAR by NMR evaluation: Principal component analysis, clustering
- Advanced wavelet algorithms
  - Noise reduction (2D and 3D)

- Wavelet based suppression of solvent signals
- Save data in wavelet format (50-80% reduced data size)

# NMRLab

NMRDAT Table 1: NMRDAT structure  
**field name**

---

NAME	Dataset name. Used for saving su
SER	Converted SER file.
MAT	Processed data matrix.
ACQUS	Acquisition parameters (3D).
PROC	Processing parameters (3D).
DISP	Display parameters.

---



---

## NMR processing functions in NMRLAB.

---

### Function

### Description

---

hft

Hilbert Transform

fft\*

Fast Fourier transform

ift\*

Inverse fast Fourier transform

dft

Fourier transform of BRUKER digital filtered data

rft

Real Fourier transform for TPPI-type data

smo

smooth = polynomial solvent filter [?]

sol

solvent filter by time-domain convolution [?]

wavewat

WAVEWAT water suppression

wdwf2

Window functions (gm, em, sine bell, cubic sine bell) © U. Günther, Eurolabs Course 2001

baseline2

Different algorithms for baseline correction



---

## Control functions in NMRLAB.

---

Function	Description
nmrlab	MATLAB script to setup parameters for NM
re	read raw data from disk
relist	read series of experiments
readser	read BRUKER ser files.
readacqus	read parameters from BRUKER ser file
snd	show data sets and sizes NMRDAT
uiphase	interactive phase correction
uicont	interactive contour plotting*
sartitr	analyze series of 2D NMR spectra (e.g. SAR
edp	edit proprocessing parameters
edd	edit display parameters
browse	browse, edit, save, export and load NMRD
xfb	process two dimensional data
xfall	process series of 2D data sets (e.g. SAR
tf	process three dimensional data
absc	2D/3D post processing baseline correction
phasend	2D/3D post processing phase correction
denoise	2D/3D post processing wavelet denoising
xyztranspose	transpose 3D structures
makespc	utility to create synthetic spectra

---

\*MATLAB  $\geq$  5.3 will automatically activate graphical tools and

---

For MATLAB 5.0-5.2 use the zoom and plottedit commands ins

---

Table 3:





[ht]

---

## Wavelet Shrinkage Parameters in NMRLAB

---

qmf_type	wavelet type	Haar, <i>Coiflet</i> ,
	(quadrature mirror filter)	<i>Daubechies</i> Vaidyanathan
par	QMF parameter	Coiflet: (3).Daubechies: 4,6,8,10 Symmlet: (8).
thr_type	Type of shrinkage	<i>hard</i> , SURE, MinMax
L	Low-Frequency cutoff for shrinkage.	must be $N = \text{number of points}$
normalize	normalize noise *	
WT_type**		<i>periodic</i> fully trans
thr	threshold value	manual <i>versal</i> = number points.

---

\*2D and 3D version of normalization has been implemented in NMRLAB. © U. Günther, EurOlab, Course 2001

---

\*\*Other wavelet transforms (i.e. the Meyer wt) are available in NMRLAB.

---

Parameters which yield good results for NMR spectroscopy have been listed in the table.

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[ht]

<b>Type of LP</b>	<b>algorithm</b>	<b>parameter ve</b>
lpsvd	SVD based linear prediction	[FID, start, stop]
prony	PRONY's method	[FID, NB, NA]
stmbr	Steiglitz McBride	[FID, NB, NA, N]
lpc	LPC	[FID, N]